

## 5.7 Inverse Filtering

The simplest approach to restoration is direct inverse filtering, where we compute an estimate,  $\hat{F}(u, v)$ , of the transform of the original image by

$$\hat{F}(u, v) = \frac{G(u, v)}{H(u, v)} \quad (5.7-1)$$

Substituting the right side of

$$G(u, v) = H(u, v)F(u, v) + N(u, v) \quad (5.1-2)$$

in (5.7-1) yields

$$\hat{F}(u, v) = F(u, v) + \frac{N(u, v)}{H(u, v)}. \quad (5.7-2)$$

The bad news is that we cannot recover the undegraded image exactly because  $N(u, v)$  is not known.

More bad news is that if the degradation function  $H(u, v)$  has zero or very small values, so the second term of (5.7-2) could easily dominate the estimate of  $\hat{F}(u, v)$ .

One approach to get around the zero or small-value problem is to limit the filter frequencies to values near the origin. As discussed earlier, we know that  $H(0, 0)$  is usually the highest value of  $H(u, v)$  in the frequency domain.

### Example 5.11: Inverse filtering

The image in Figure 5.25 (b) was inverse filtered with

$$\hat{F}(u, v) = \frac{G(u, v)}{H(u, v)} \quad (5.7-1)$$

using the exact **inverse** of the **degradation function** that generated that image. That is, the **degradation function** used was

$$H(u, v) = e^{-k[(u-M/2)^2 + (v-N/2)^2]^{5/6}}$$

with  $k = 0.0025$ . In this case,  $M = N = 480$ .

Although a **Gaussian-shape function** has no zeros and it is not a concern here, the **degradation** values become so small that the result of **full inverse filtering** shown in Figure 5.27 (a) is useless.

a	b
c	d

**FIGURE 5.27**

Restoring Fig. 5.25(b) with Eq. (5.7-1). (a) Result of using the full filter. (b) Result with  $H$  cut off outside a radius of 40; (c) outside a radius of 70; and (d) outside a radius of 85.



Figure 5.27 (b) through (d) show the results of cutting off values of the ratio  $G(u, v) / H(u, v)$  outside a radius of 40, 70, and 85, respectively.

Values above 70 started to produce degraded images, and further increases in radius values would produce images that looked more and more like Figure 5.27 (a).

The results in Example 5.11 show the poor performance of direct inverse filtering in general.

## 5.8 Minimum Mean Square Error (Wiener) Filtering

Here we discuss an approach that incorporates both the **degradation function** and **statistical characteristics** of noise into the **restoration process**.

Considering images and noise as **random variables**, the objective is to find an estimate  $\hat{f}$  of the uncorrupted image  $f$  such that the **mean square error** between them is minimized.

The error measure is given by

$$e^2 = E\{(f - \hat{f})^2\} \quad (5.8-1)$$

where  $E\{\bullet\}$  is the expected value of the argument.

By assuming that

1. the noise and the image are **uncorrelated**;
2. one or the other has **zero mean**;
3. the intensity levels in the estimate are a **linear function** of the levels in the degraded image.

Then, the minimum of the error function in (5.8-1) is given in the **frequency domain** by the expression

$$\begin{aligned} \hat{F}(u, v) &= \left[ \frac{H^*(u, v)S_f(u, v)}{S_f(u, v)|H(u, v)|^2 + S_\eta(u, v)} \right] G(u, v) \\ &= \left[ \frac{H^*(u, v)}{|H(u, v)|^2 + S_\eta(u, v) / S_f(u, v)} \right] G(u, v) \\ &= \left[ \frac{1}{H(u, v)} \frac{|H(u, v)|^2}{|H(u, v)|^2 + S_\eta(u, v) / S_f(u, v)} \right] G(u, v) \end{aligned} \quad (5.8-2)$$

The terms in (5.8-2) are as follows:

$\hat{F}(u, v)$  is the frequency domain estimate

$G(u, v)$  is the transform of the degraded image

$H(u, v)$  is the transform of the degradation function

$H^*(u, v)$  is complex conjugate of  $H(u, v)$

$$|H(u, v)|^2 = H^*(u, v)H(u, v)$$

$S_\eta(u, v) = |N(u, v)|^2 =$  power spectrum of the noise

$S_f(u, v) = |F(u, v)|^2 =$  power spectrum of the undegraded image

This result is known as the Wiener filter, which also is commonly referred to as the minimum mean square error filter or the least square error filter.

The Wiener filter does not have the same problem as the inverse filter with zeros in the degradation function, unless the entire denominator is zero for the same value(s) of  $u$  and  $v$ .

If the noise is zero, then the Wiener filter reduces to the inverse filter.

One of the most important measures is the signal-to-noise ratio, approximated using frequency domain quantities such as

$$SNR = \frac{\sum_{u=0}^{M-1} \sum_{v=0}^{N-1} |F(u, v)|^2}{\sum_{u=0}^{M-1} \sum_{v=0}^{N-1} |N(u, v)|^2} \quad (5.8-3)$$

The **mean square error** given in statistical form in (5.8-1) can be approximated also in terms a summation involving the original and restored images:

$$MSE = \frac{1}{MN} \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} [f(x,y) - \hat{f}(x,y)]^2 \quad (5.8-4)$$

If one considers the restored image to be **signal** and the **difference** between this image and the original to be **noise**, we can define a **signal-to-noise** ratio in the **spatial domain** as

$$SNR = \frac{\sum_{x=0}^{M-1} \sum_{y=0}^{N-1} \hat{f}(x,y)^2}{\sum_{x=0}^{M-1} \sum_{y=0}^{N-1} [f(x,y) - \hat{f}(x,y)]^2} \quad (5.8-5)$$

The closer  $f$  and  $\hat{f}$  are, the larger this ratio will be.

If we are dealing with **white noise**, the **spectrum**  $|N(u,v)|^2$  is a constant, which simplifies things considerably. However,  $|F(u,v)|^2$  is usually unknown.

An approach is used frequently when these quantities are not known or cannot be estimated:

$$\hat{F}(u,v) = \left[ \frac{1}{H(u,v)} \frac{|H(u,v)|^2}{|H(u,v)|^2 + K} \right] G(u,v) \quad (5.8-6)$$

where  $K$  is a specified constant that is added to all terms of  $|H(u,v)|^2$ .

Note: **White noise** is a random **signal** (or process) with a flat **power spectral density**. In other words, the signal contains equal power within a fixed **bandwidth** at any center frequency.

### Example 5.12: Comparison of inverse and Wiener filtering

Figure 5.28 shows the advantage of Wiener filtering over direct inverse filtering.



a b c

**FIGURE 5.28** Comparison of inverse and Wiener filtering. (a) Result of full inverse filtering of Fig. 5.25(b). (b) Radially limited inverse filter result. (c) Wiener filter result.

Figure 5.28 (a) is the full inverse-filtered result from Figure 5.27 (a).

Figure 5.28 (b) is the radially limited inverse result of Figure 5.27 (c).

Figure 5.28 (c) shows the result obtained using

$$\hat{F}(u, v) = \left[ \frac{1}{H(u, v)} \frac{|H(u, v)|^2}{|H(u, v)|^2 + K} \right] G(u, v) \quad (5.8-6)$$

with the degradation function

$$H(u, v) = e^{-k[(u-M/2)^2 + (v-N/2)^2]^{5/6}}$$

used in Example 5.11. The value of  $K$  was chosen interactively to yield the best visual result.

By comparing Figure 5.25 (a) and Figure 5.28 (c), we see that the Wiener filter yielded a result very close in appearance to the original image.



### Example 5.13: Further comparisons of Wiener filtering



**FIGURE 5.29** (a) 8-bit image corrupted by motion blur and additive noise. (b) Result of inverse filtering. (c) Result of Wiener filtering. (d)–(f) Same sequence, but with noise variance one order of magnitude less. (g)–(i) Same sequence, but noise variance reduced by five orders of magnitude from (a). Note in (h) how the deblurred image is quite visible through a “curtain” of noise.



## 5.10 Geometric Mean Filter

It is possible to generalize the **Wiener filter** slightly to the so-called **geometric mean filter**:

$$\hat{F}(u, v) = \left[ \frac{H^*(u, v)}{|H(u, v)|^2} \right]^\alpha \left[ \frac{H^*(u, v)}{|H(u, v)|^2 + \beta \left[ \frac{S_\eta(u, v)}{S_f(u, v)} \right]} \right]^{1-\alpha} G(u, v), \quad (5.10-1)$$

where  $\alpha$  and  $\beta$  are positive real constants.

If  $\alpha = 1$ , this filter reduces to the **inverse filter**.

If  $\alpha = 0$ , the filter becomes the so-called **parametric Wiener filter**, which reduces to the standard **Wiener filter** when  $\beta = 1$ .

If  $\alpha = 1/2$ , this filter becomes a product of the two quantities raised to the same power, which is the definition of the **geometric mean**. When  $\beta = 1$ , the filter is also commonly referred to as the **spectrum equalization filter**.

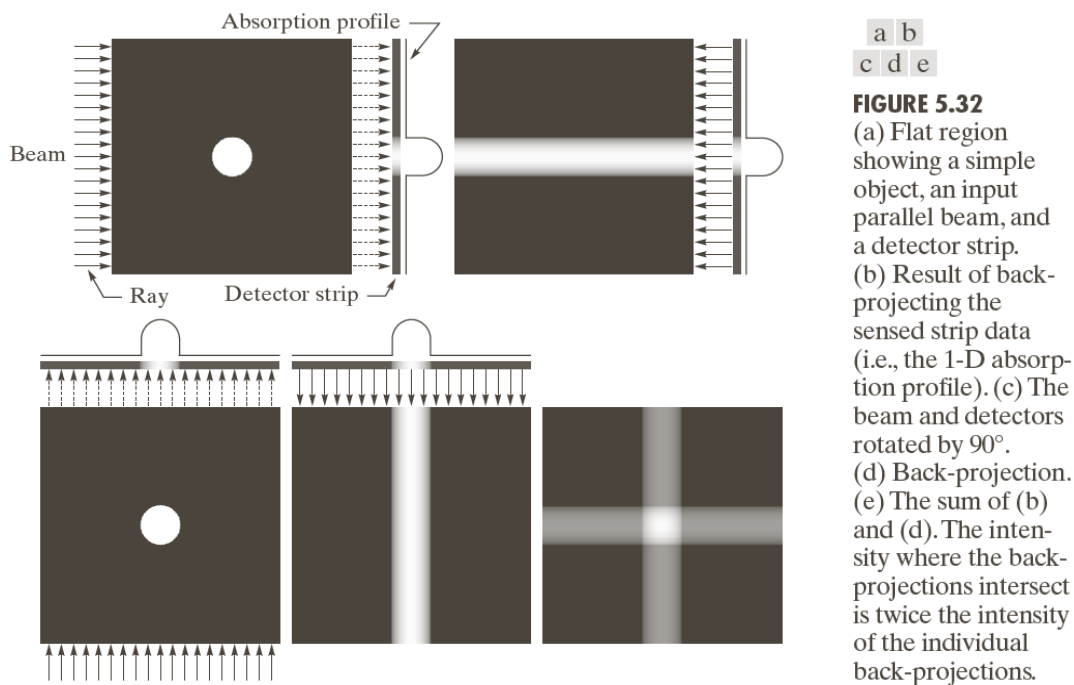
With  $\beta = 1$ , as  $\alpha$  decreases below  $1/2$ , the filter performance will tend more toward to the **inverse filter**; as  $\alpha$  increases above  $1/2$ , the filter will behave more like the **Wiener filter**.

## 5.11 Image Reconstruction from Projections

In this section, we will examine the problem of reconstructing an image from a series of **projections**, with a focus on **X-ray computed tomography (CT)**, which is one of the principal applications of digital image processing in medicine.

### Introduction

Consider Figure 5.32 (a), which consists of a single object on a **uniform background**.



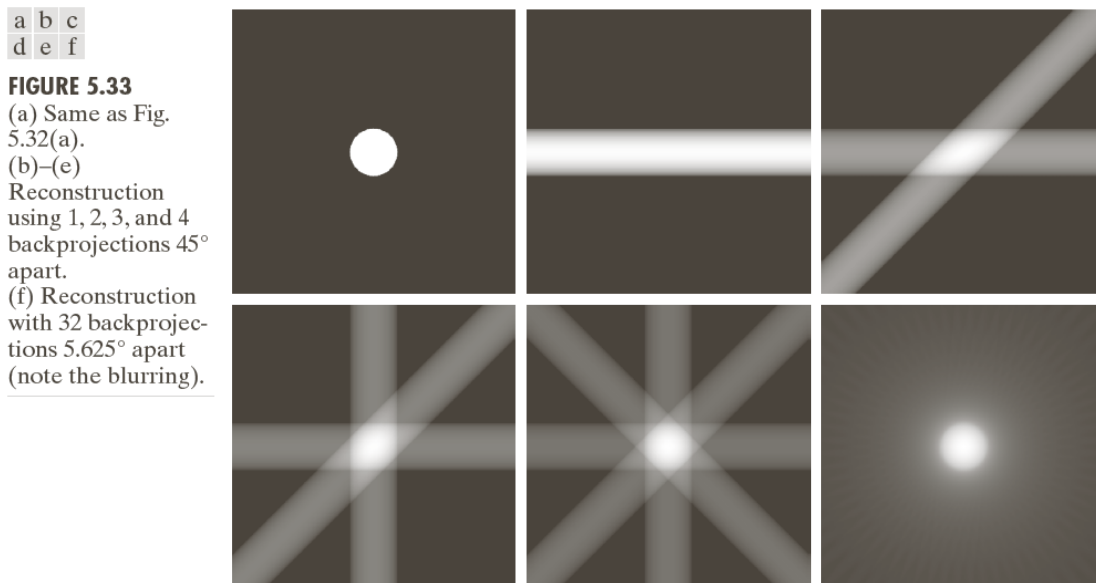
Suppose that we pass a thin, flat beam of X-rays from left to right, and assume that the energy of the beam is absorbed more by the object than by the background. Using a strip of X-ray absorption detectors on the other side will yield the signal, whose amplitude (intensity) is proportional to absorption.

The approach is to project the 1-D signal back across the direction from which the beam came, as Figure 5.32 (b) shows. This approach is called **backprojection**.

We certainly cannot determine a single object or a multitude of objects along the path of the beam by a single project.

If we rotate the position of the source-detector pair by  $90^\circ$  and repeat the previous procedure, we will get a **backprojection** image shown in **Figure 5.32 (d)**. Adding this result to **Figure 5.32 (b)** will result an image illustrated in **Figure 5.32 (e)**.

We should be able to learn more about the shape of the object in question by taking more views in the same manner, as shown in **Figure 5.33**.



As the number of projections increases, the strength of **non-intersecting backprojects** decreases relative to the strength of regions in which multiple **backprojects** intersect.

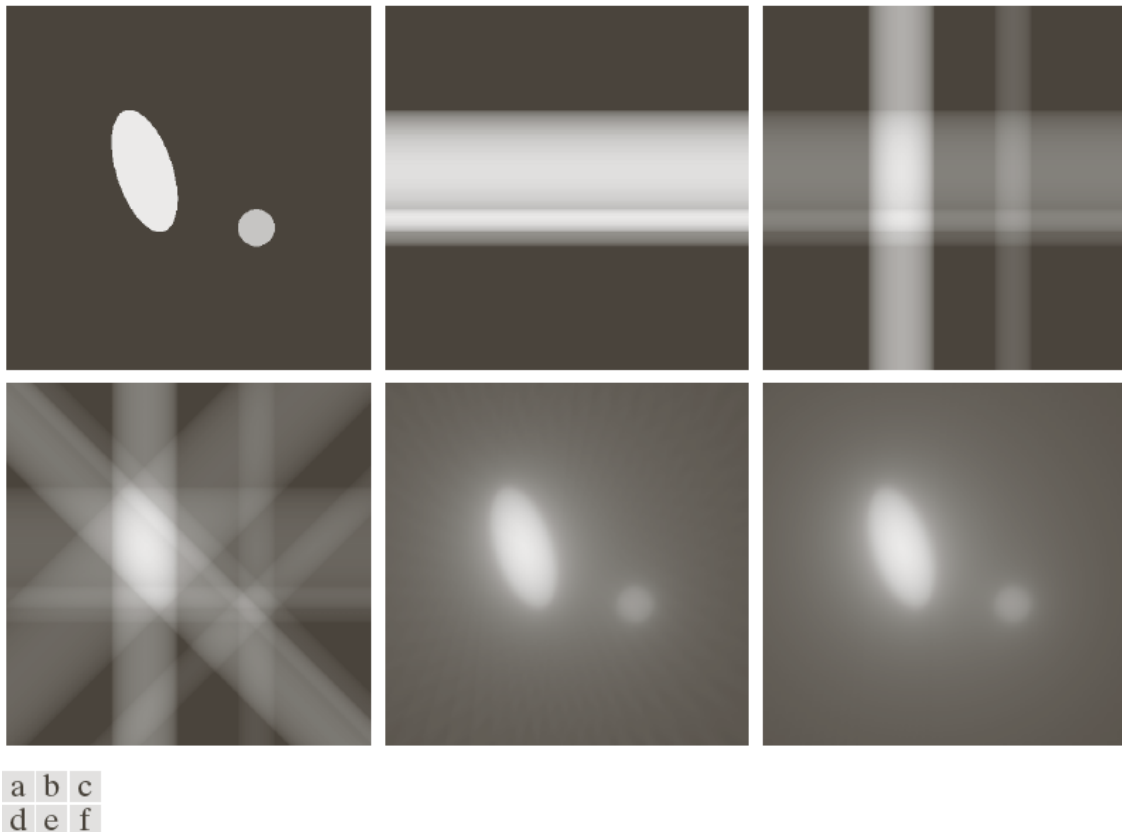
**Figure 5.33 (f)** shows the result formed from **32 projections**.

The reconstructed image seems to be a reasonably good approximation to the shape of the original object. However, the image is **blurred** by a “**halo**” effect, which shows a “star” in **Figure 5.33 (e)**. As the number of views increases, the shape of the “**halo**” becomes circular, as shown in **Figure 5.33 (f)**.

**Blurring** in **CT** reconstruction is an important issue and will be addressed in later discussion.

Since the projections  $180^\circ$  apart are **mirror images** of each other, we only need to consider angle increments halfway around a circle in order to generate all the projects required for reconstruction.

**Example 5.16: Backprojection of a simple planar region containing two objects**



**FIGURE 5.34** (a) A region with two objects. (b)–(d) Reconstruction using 1, 2, and 4 backprojections  $45^\circ$  apart. (e) Reconstruction with 32 backprojections  $5.625^\circ$  apart. (f) Reconstruction with 64 backprojections  $2.8125^\circ$  apart.

## Principles of Computed Tomography (CT)

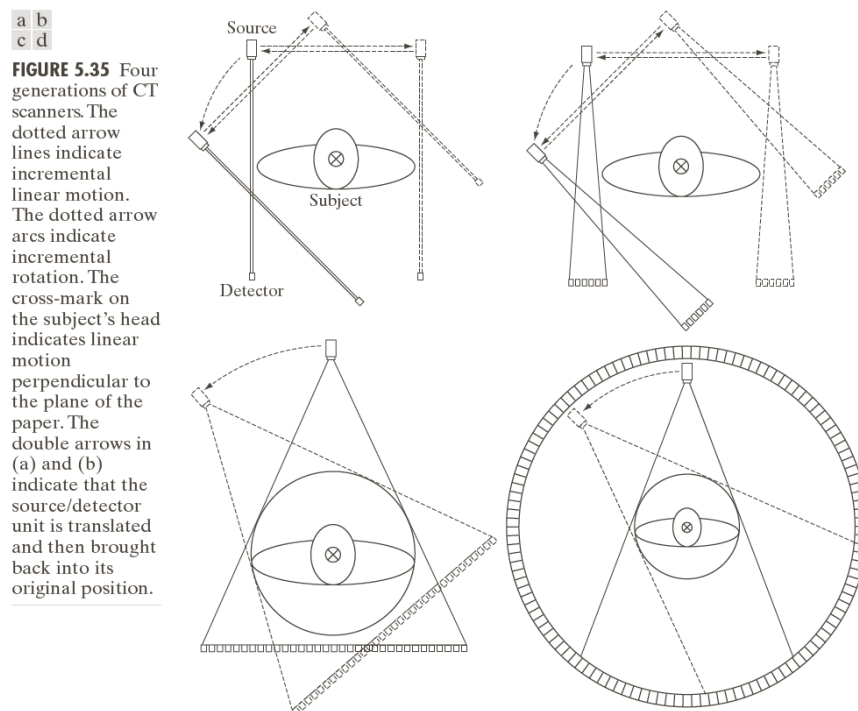
The theoretical foundation of CT dates back to [Johann Radon](#), a mathematician from Vienna who derived a method in 1907 for projecting a 2-D object along parallel rays as part of his work on line integrals. The method is referred as the [Radon transform](#) now.

[Allan M. Cormack](#), a physicist at Tufts University, partially “rediscovered” these concepts and applied them to CT. [Cormack](#) published his initial findings in 1963 and 1964. He provided the mathematical formulae needed for the reconstruction and built a CT prototype to show his ideas.

Working independently, electrical engineer [Godfrey N. Hounsfield](#) and his colleagues at EMI in London formulated a similar solution and built the first medical CT machine.

[Cormack](#) and [Hounsfield](#) shared the [1979 Nobel Prize in Medicine](#) for their contributions to medical tomography.

[Figure 5.35](#) shows the first four generations of CT scanners.



The fifth-generation (G5) CT scanners eliminate all mechanical motion by employing electron beams controlled electromagnetically.

The sixth-generation (G6) CT scanners rotate the source-detector pair continuously through  $360^\circ$ , while the patient is moved at a constant speed along the axis perpendicular to the scan.

The seventh-generation (G7) CT scanners (also called multislice CT scanners) use parallel banks of detectors to collect volumetric CT data simultaneously.

### Projections and the Radon Transform

A straight line in Cartesian coordinates can be described either by its slope-intercept form

$$y = ax + b,$$

or, as in Figure 5.36, by its normal representation

$$x \cos \theta + y \sin \theta = \rho. \quad (5.11-1)$$

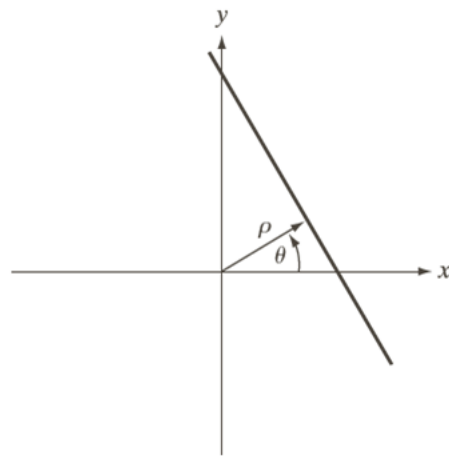
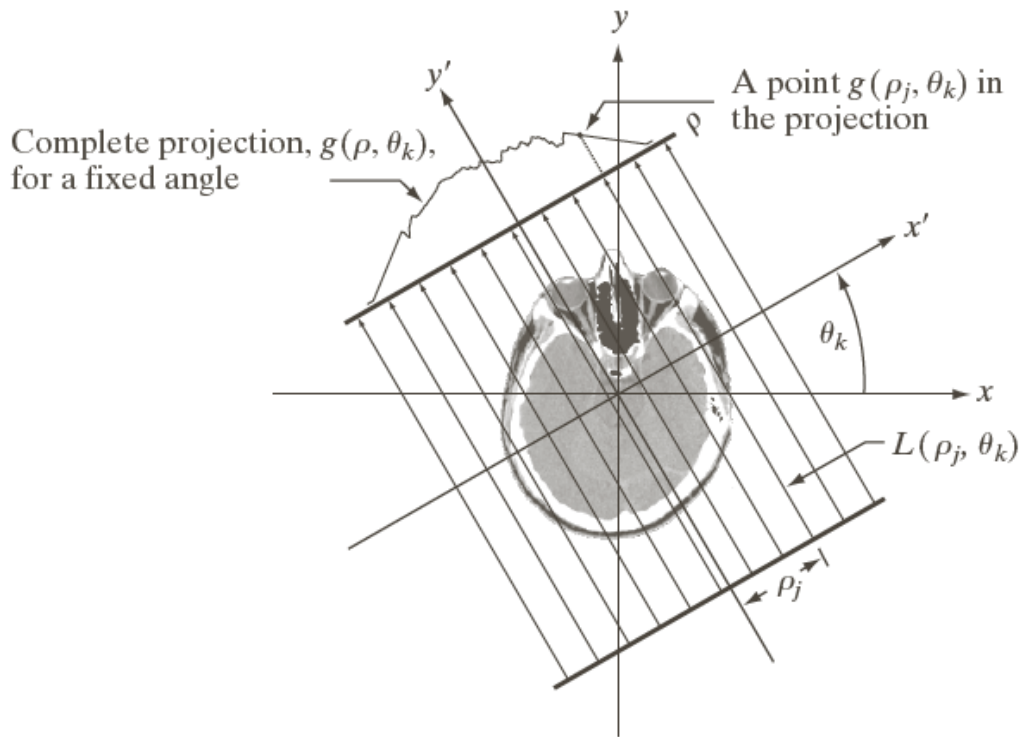


Figure 5.36 Normal representation of a straight line.

The projection of a **parallel-ray beam** may be modeled by a set of such lines, as shown in **Figure 5.37**.



**Figure 5.37** Geometry of a parallel-ray beam.

An arbitrary point in the projection signal is given by the raysum along the line

$$x \cos \theta_k + y \sin \theta_k = \rho_j .$$

In the case of continuous, the raysum is a **line integral**, given by

$$g(\rho_j, \theta_k) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} f(x, y) \delta(x \cos \theta_k + y \sin \theta_k - \rho_j) dx dy \quad (5.11-2)$$

Recall the properties of the **impulse**,  $\delta$ , the right side of (5.11-2) is zero unless the argument of  $\delta$  is zero. It indicates that the integral is computed only along the line  $x \cos \theta_k + y \sin \theta_k = \rho_j$ .



If we consider all values of  $\rho$  and  $\theta$ , (5.11-2) generalizes

$$g(\rho, \theta) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} f(x, y) \delta(x \cos \theta + y \sin \theta - \rho) dx dy . \quad (5.11-3)$$

The equation (5.11-3) gives the **projection** of  $f(x, y)$  along an arbitrary line in the  $xy$ -plane, is called the **Radon transform**.

The **Radon transform** is the cornerstone of reconstruction from projections, with **CT** being its principle application in the field of image processing.

In the discrete case, (5.11-3) becomes

$$g(\rho, \theta) = \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} f(x, y) \delta(x \cos \theta + y \sin \theta - \rho) , \quad (5.11-4)$$

where  $x$ ,  $y$ ,  $\rho$ , and  $\theta$  are now **discrete** variables.

If we fix  $\theta$  and allow  $\rho$  to vary, (5.11-4) simply sums the pixels of  $f(x, y)$  along the line defined by specified values of these two parameters.

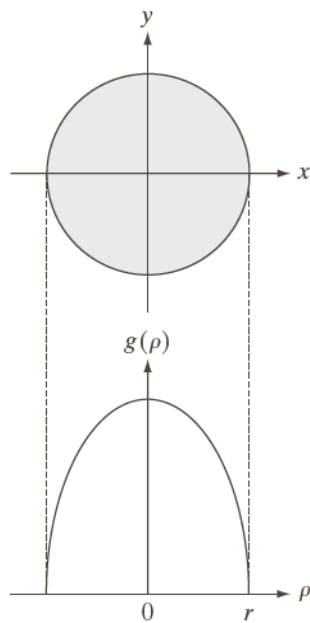
Incrementing through all values of  $\rho$  required to span the image (with  $\theta$  fixed) yields one projection. Changing  $\theta$  and repeating the same procedure will yield another projection.

### Example 5.17: Using the Radon transform to obtain the projection of a circular region.

We want to obtain the **Radon transform** for the projection of the circular object

$$f(x, y) = \begin{cases} A & x^2 + y^2 \leq r^2 \\ 0 & \text{otherwise} \end{cases},$$

where  $A$  is a constant and  $r$  is the radius of the object. The circular object is shown in **Figure 5.38 (a)**.



**FIGURE 5.38** A disk and a plot of its Radon transform, derived analytically. Here we were able to plot the transform because it depends only on one variable. When  $g$  depends on both  $\rho$  and  $\theta$ , the Radon transform becomes an image whose axes are  $\rho$  and  $\theta$ , and the intensity of a pixel is proportional to the value of  $g$  at the location of that pixel.

Since the object is circularly symmetric, its projections are the same for all angles, so all we need is to obtain the projection for  $\theta = 0^\circ$ . From (5.11-3), we get

$$\begin{aligned} g(\rho, \theta) &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} f(x, y) \delta(x - \rho) dx dy \\ &= \int_{-\infty}^{\infty} f(\rho, y) dy \end{aligned}$$

This is a **line integral** along the line  $L(\rho, 0)$ .

Note that  $g(\rho, \theta) = 0$  when  $|\rho| > r$ . When  $|\rho| \leq r$ , the integral is evaluated from  $y = -\sqrt{r^2 - \rho^2}$  to  $y = \sqrt{r^2 - \rho^2}$ .

Therefore,

$$g(\rho, \theta) = \int_{-\sqrt{r^2 - \rho^2}}^{\sqrt{r^2 - \rho^2}} f(\rho, y) dy = \int_{-\sqrt{r^2 - \rho^2}}^{\sqrt{r^2 - \rho^2}} A dy$$

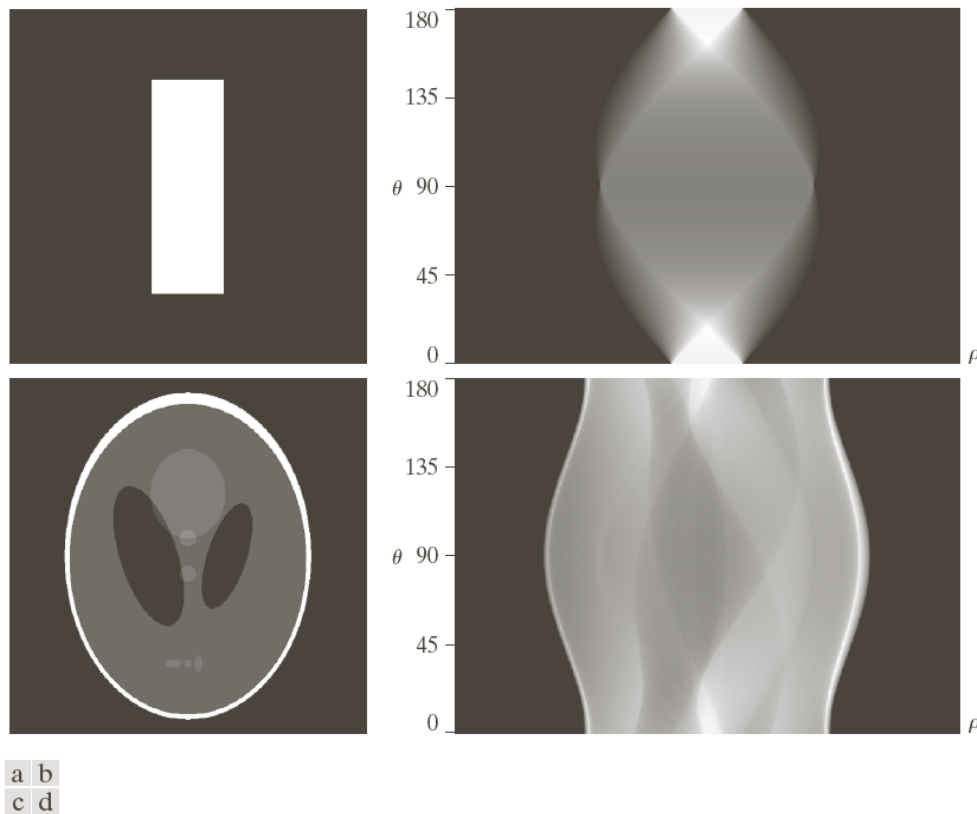
It yields

$$g(\rho, \theta) = g(\rho) = \begin{cases} 2A\sqrt{r^2 - \rho^2} & |\rho| \leq r \\ 0 & \text{otherwise} \end{cases}$$

Figure 5.38 (b) shows the result.

$g(\rho, \theta) = g(\rho)$  indicates that  $g$  is independent of  $\theta$  because the object is symmetric about the origin.

When the **Radon transform**,  $g(\rho, \theta)$ , is displayed as an image with  $\rho$  and  $\theta$  as rectilinear coordinates, the result is called a **sinogram**, similar in concept to displaying the **Fourier spectrum**. Like the **Fourier spectrum**, a **sinogram** contains the data necessary to reconstruct  $f(x, y)$ .



**FIGURE 5.39** Two images and their sinograms (Radon transforms). Each row of a sinogram is a projection along the corresponding angle on the vertical axis. Image (c) is called the *Shepp-Logan phantom*. In its original form, the contrast of the phantom is quite low. It is shown enhanced here to facilitate viewing.

Figure 5.39 (b) is the **sinogram** of the rectangle shown in Figure 5.39 (a).

Figure 5.39 (c) shows an image of the **Shepp-Logan phantom**, a widely used synthetic image designed to simulate the absorption of major areas of the brain. The **sinogram** of Figure 5.39 (c) is shown in Figure 5.39 (d).

To obtain a formal expression for a **back-projected** image from **Radon transform**, referring to **Figure 5.37**, we begin with a **single** point,  $g(\rho_j, \theta_k)$ , of the complete projection,  $g(\rho, \theta_k)$ , for a **fixed** value of rotation,  $\theta_k$ .

Forming part of an image by **back-projecting** this single point is simply to copy the line  $L(\rho_j, \theta_k)$  onto the image, where the value of each point in that line is  $g(\rho_j, \theta_k)$ . Repeating this process of all values of  $\rho_j$  in the projected signal results

$$f_{\theta_k}(x, y) = g(\rho, \theta_k) = g(x \cos \theta_k + y \sin \theta_k, \theta_k).$$

This equation holds for an arbitrary value of  $\theta_k$ , therefore, we can write in general that the image formed from a single **backprojection** obtained at an angle  $\theta$  is given by

$$f_{\theta}(x, y) = g(x \cos \theta + y \sin \theta, \theta). \quad (5.11-5)$$

We form the final image by integrating over all the **back-projected** images

$$f(x, y) = \int_0^{\pi} f_{\theta}(x, y) d\theta \quad (5.11-6)$$

In the **discrete** case, the integral becomes a sum of all **back-projected** images:

$$f(x, y) = \sum_{\theta=0}^{\pi} f_{\theta}(x, y) \quad (5.11-7)$$

For example, if  $0.5^\circ$  increments are being used, the summation is from 0 to 179.5.

A **back-projected** image formed in this manner is referred to as a **laminogram**, which is only an approximation to the image from which the projections were generated.

### Example 5.18: Obtaining back-projected images from sinograms

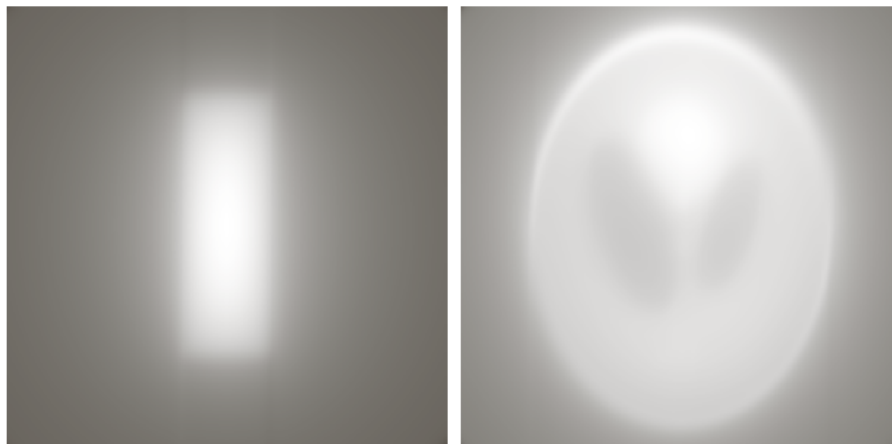
Equation

$$f(x, y) = \sum_{\theta=0}^{\pi} f_{\theta}(x, y) \quad (5.11-7)$$

was used to generate the **back-projected** images in Figure 5.32 through Figure 5.34 from projections obtained with

$$g(\rho, \theta) = \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} f(x, y) \delta(x \cos \theta + y \sin \theta - \rho). \quad (5.11-4)$$

These equations were also used to generate Figure 5.40 (a) and Figure 5.40 (b), which show the **back-projected** images corresponding to the **sinograms** in Figure 5.39 (b) and Figure 5.39 (d).



a b

**FIGURE 5.40**  
Backprojections  
of the sinograms  
in Fig. 5.39.

Note that there is a significant amount of **blurring** shown in Figure 5.40 (a) and (b). It is obvious that a straight use of Equations (5.11-4) and (5.11-7) will not yield acceptable results.

## The Fourier-Slice Theorem

The relationship relating the 1-D Fourier transform of a projection and the 2-D Fourier transform of the region from which the projection was obtained is the basis for reconstruction methods capable of dealing with the blurring problem.

The 1-D Fourier transform of a projection with respect to  $\rho$  is

$$G(\omega, \theta) = \int_{-\infty}^{\infty} g(\rho, \theta) e^{-j2\pi\omega\rho} d\rho \quad (5.11-8)$$

where  $\omega$  is the frequency variable, and this expression is for a given value of  $\theta$ .

Substituting

$$g(\rho, \theta) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} f(x, y) \delta(x \cos \theta + y \sin \theta - \rho) dx dy \quad (5.11-3)$$

for  $g(\rho, \theta)$  results the expression

$$\begin{aligned} G(\omega, \theta) &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} f(x, y) \delta(x \cos \theta + y \sin \theta - \rho) e^{-j2\pi\omega\rho} dx dy d\rho \\ &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} f(x, y) \left[ \int_{-\infty}^{\infty} \delta(x \cos \theta + y \sin \theta - \rho) e^{-j2\pi\omega\rho} d\rho \right] dx dy \\ &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} f(x, y) e^{-j2\pi\omega(x \cos \theta + y \sin \theta)} dx dy \quad (5.11-9) \end{aligned}$$

By letting  $u = \omega \cos \theta$  and  $v = \omega \sin \theta$ , (5.11-9) becomes

$$G(\omega, \theta) = \left[ \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} f(x, y) e^{-j2\pi(ux + vy)} dx dy \right]_{u=\omega \cos \theta; v=\omega \sin \theta} \quad (5.11-10)$$

We recognize (5.11-10) as the 2-D Fourier transform of  $f(x, y)$  evaluated at the values of  $u$  and  $v$  indicated.



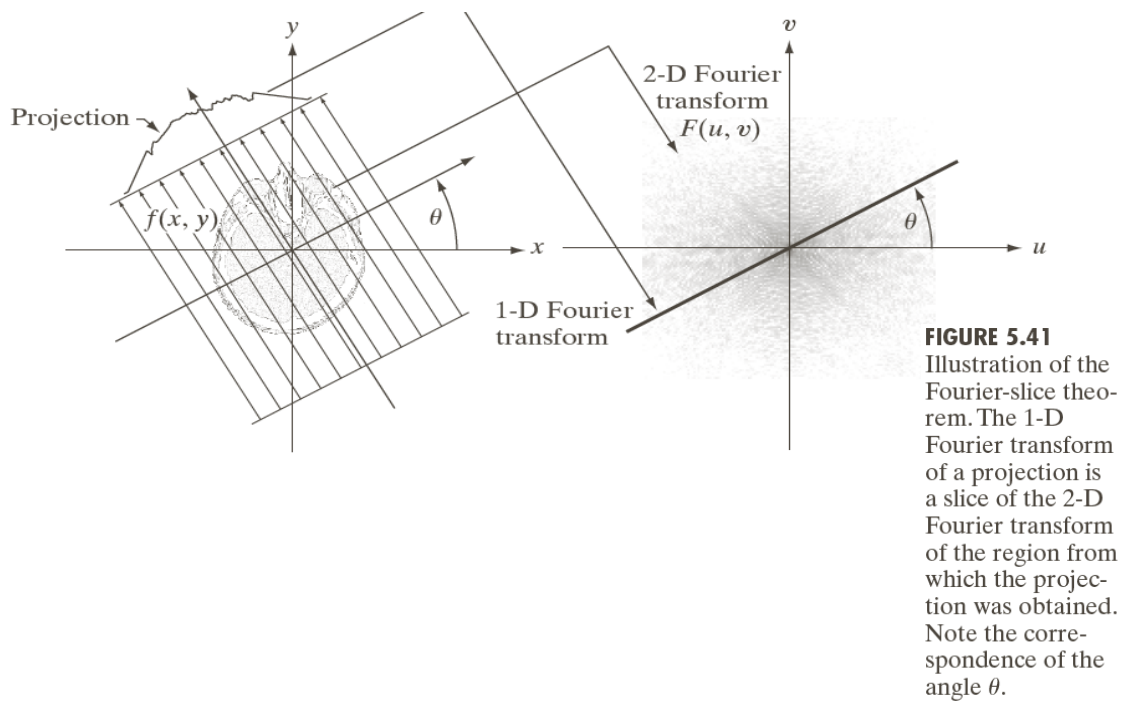
Equation (5.11-10) leads to

$$G(\omega, \theta) = [F(u, v)]_{u=\omega \cos \theta; v=\omega \sin \theta} = F(\omega \cos \theta, \omega \sin \theta), \quad (5.11-11)$$

which is known as the **Fourier-slice theorem** (or the **projection-slice theorem**).

The **Fourier-slice theorem** states that the **Fourier transform** of a **projection** is a **slice** of the **2-D Fourier transform** of the **region** from which the projection was obtained.

This terminology can be explained with **Figure 5.41**.



As **Figure 5.41** shows, the **1-D Fourier transform** of an arbitrary projection is obtained by extracting the values of  $F(u, v)$  along a line oriented at the same angle as the angle used in generating the projection.

In principle, we could obtain  $f(x, y)$  simply by obtaining the **inverse Fourier transform**  $F(u, v)$ , though it is expensive computationally with the involvement of inverting a **2-D transform**.

## Reconstruction Using Parallel-Beam Filtered Backprojections

Regarding to the **blurred** results, fortunately, there is a simple solution based on filtering the projections before computing the **backprojections**.

Recall

$$f(t, z) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} F(\mu, \nu) e^{j2\pi(\mu t + \nu z)} d\mu d\nu, \quad (4.5-8)$$

the 2-D inverse Fourier transform of  $F(u, v)$  is

$$f(x, y) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} F(u, v) e^{j2\pi(ux + vy)} du dv. \quad (5.11-12)$$

As in (5.11-10) and (5.11-11), letting  $u = \omega \cos \theta$  and  $v = \omega \sin \theta$ , we can express (5.11-12) in polar coordinates:

$$f(x, y) = \int_0^{2\pi} \int_0^{\infty} F(\omega \cos \theta, \omega \sin \theta) e^{j2\pi\omega(x \cos \theta + y \sin \theta)} \omega d\omega d\theta \quad (5.11-13)$$

Then, using the **Fourier-slice theorem**, we have

$$f(x, y) = \int_0^{2\pi} \int_0^{\infty} G(\omega, \theta) e^{j2\pi\omega(x \cos \theta + y \sin \theta)} \omega d\omega d\theta. \quad (5.11-14)$$

Using the fact that  $G(\omega, \theta + \pi) = G(-\omega, \theta)$ , we can express (5.11-14) as

$$f(x, y) = \int_0^{\pi} \int_{-\infty}^{\infty} |\omega| G(\omega, \theta) e^{j2\pi\omega(x \cos \theta + y \sin \theta)} d\omega d\theta. \quad (5.11-15)$$

In terms of integration with respect to  $|\omega|$ , the term  $x \cos \theta + y \sin \theta$  is a constant, which is recognized as  $\rho$ . Thus, (5.11-15) can be written as

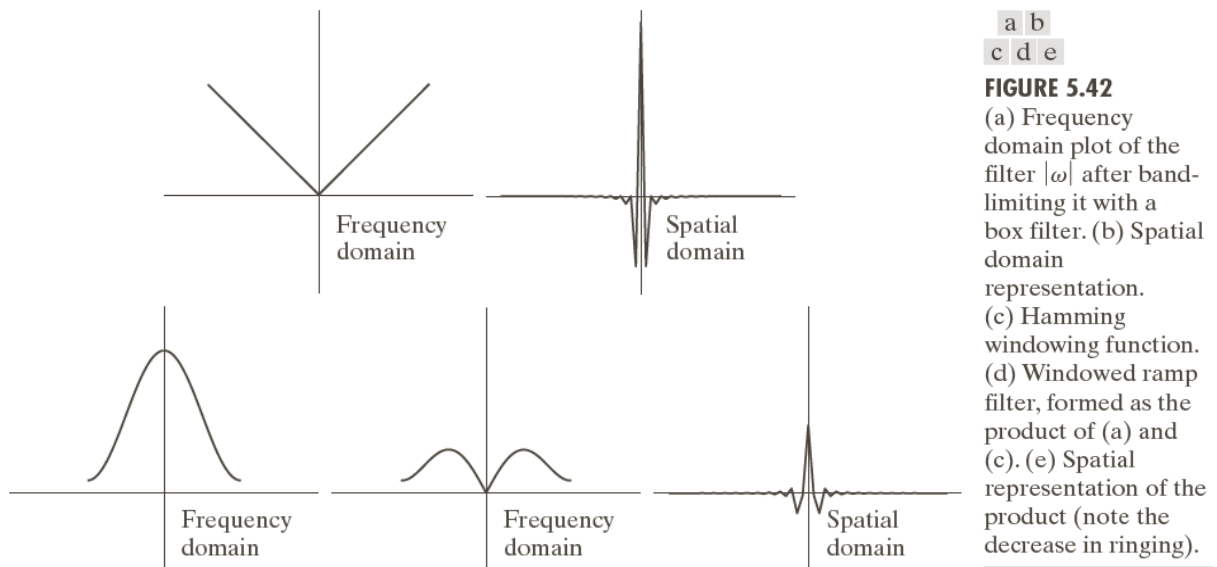
$$f(x, y) = \int_0^{\pi} \left[ \int_{-\infty}^{\infty} |\omega| G(\omega, \theta) e^{j2\pi\omega\rho} d\omega \right]_{\rho=x \cos \theta + y \sin \theta} d\theta. \quad (5.11-16)$$

Recall

$$f(t) = \int_{-\infty}^{\infty} F(\mu) e^{j2\pi\mu t} d\mu, \quad (4.2-17)$$

the inner expression in (5.11-16) is a 1-D inverse Fourier transform with the added term  $|\omega|$ .

Based on the discussion in Section 4.7,  $|\omega|$  is a one-dimensional filter function.



**FIGURE 5.42**

(a) Frequency domain plot of the filter  $|\omega|$  after band-limiting it with a box filter. (b) Spatial domain representation. (c) Hamming windowing function. (d) Windowed ramp filter, formed as the product of (a) and (c). (e) Spatial representation of the product (note the decrease in ringing).

$|\omega|$  is not integrable, because its amplitude extends to  $+\infty$  in both directions, so the inverse Fourier transform is undefined.

In practice, the approach is to window the ramp so it becomes zero outside of defined frequency interval, as shown in Figure 5.42 (a).

Figure 5.42 (b) shows its spatial domain representation, obtained by computing its inverse Fourier transform. The resulting windowed filter exhibits noticeable ringing in the spatial domain. As discussed in Chapter 4, windowing with a smooth function will help in this situation.

An **M-point discrete window function** used frequently for implementation with the **1-D FFT** is given by

$$h(\omega) = \begin{cases} c + (c - 1) \cos \frac{2\pi\omega}{M - 1} & 0 \leq \omega \leq (M - 1) \\ 0 & \text{otherwise} \end{cases} \quad (5.11-17)$$

When  $c = 0.54$ , this function is called the **Hamming window**.

Figure 5.42 (c) is a plot of the **Hamming window**, and Figure 5.42 (d) shows the product of this window and the **band-limited ramp filter** shown in Figure 5.42 (a).

Figure 5.42 (e) shows the representation of the product in the **spatial domain**, obtained by computing the **inverse FFT**.

Comparing Figure 5.42 (e) and Figure 5.42 (b), we can find that ringing was reduced in the window ramp.

On the other hand, because the width of the central lobe in Figure 5.42 (e) is slightly wider than that of Figure 5.42 (b), we would expect backprojections based on a **Hamming window** to have **less ringing** but be slightly **more blurred**.

Recalling

$$G(\omega, \theta) = \int_{-\infty}^{\infty} g(\rho, \theta) e^{-j2\pi\omega\rho} d\rho \quad (5.11-8)$$

that  $G(\omega, \theta)$  is the **1-D Fourier transform** of  $g(\rho, \theta)$ , which is a single projection obtained at a fixed angle,  $\theta$ .

Equation

$$f(x, y) = \int_0^\pi \left[ \int_{-\infty}^{\infty} |\omega| G(\omega, \theta) e^{j2\pi\omega\rho} d\omega \right]_{\rho=x\cos\theta+y\sin\theta} d\theta \quad (5.11-16)$$

states that the complete, **back-projected** image  $f(x, y)$  is obtained as follows:

1. Compute the **1-D Fourier transform** of **each projection**.
2. Multiply each **Fourier transform** by the filter function  $|\omega|$ , which has been multiplied by a suitable (e.g., **Hamming window**).
3. Obtain the **inverse 1-D Fourier transform** of each resulting filtered transform.
4. Integrate (sum) all the **1-D inverse transform** from **Step 3**.

This image reconstruction approach is called **filtered backprojection**.

In practice, because the data are **discrete**, all **frequency domain** computations are carried out using a **1-D FFT** algorithm, and filtering is implemented using the same basic procedure explained in **Chapter 4** for 2-D functions.

### Example 5.19: Image reconstruction using filtered backprojections

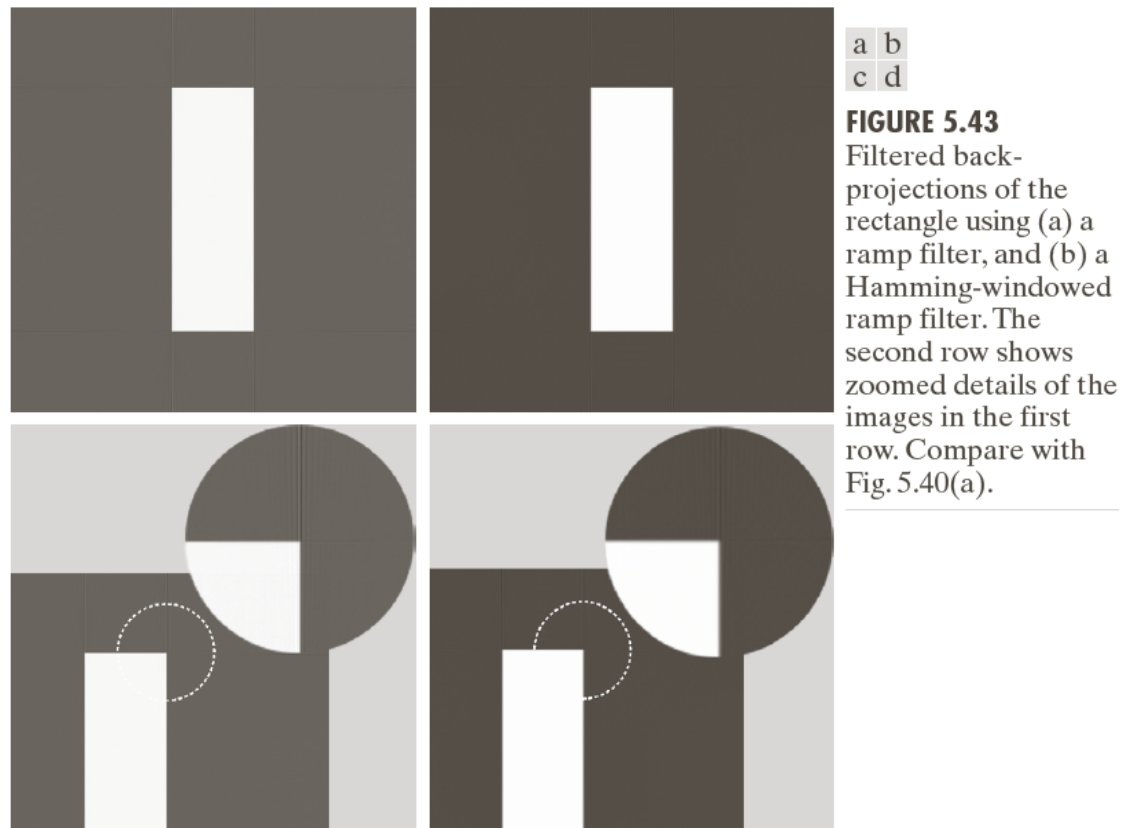
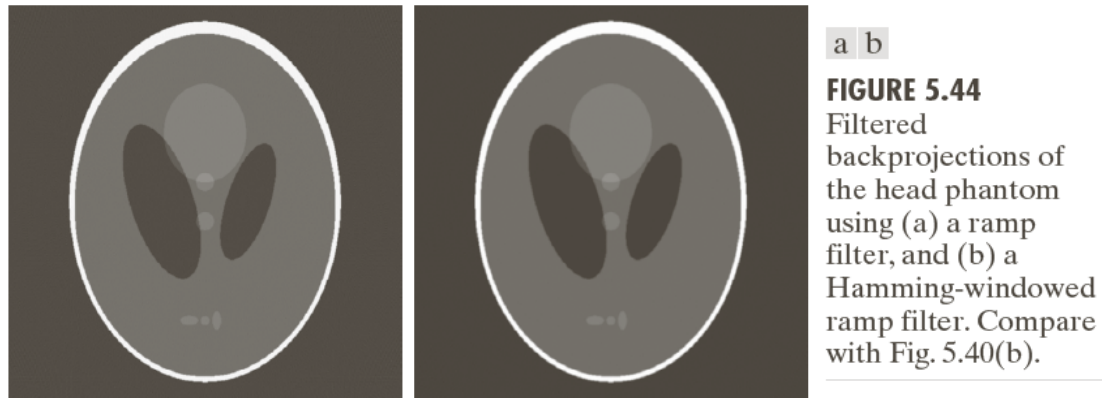


Figure 5.43 (a) shows the rectangle reconstructed using a **ramp filter**. The most vivid feature of this result is the absence of any visually detectable **blurring**. However, **ringing** is present, visible as faint lines, especially around the corners of the rectangle. Figure 5.43 (c) can show these lines in the zoomed section.

Using a **Hamming window** on the **ramp filter** helped considerably with the **ringing** problem, at the expense of slight blurring, as Figure 5.43 (b) and Figure 5.43 (d) show.



The reconstructed phantom images shown in Figure 5.44 are from using the un-windowed ramp filter and a Hamming window on the ramp filter.

Since the phantom image does not have transitions that are sharp and prominent as the rectangle, so ringing is imperceptible in this case, though result shown in Figure 5.44 (b) is a slightly smooth than that of Figure 5.44 (a).

The discussion has been based on obtaining filtered backprojections via an FFT implementation. However, from the convolution theorem introduced in Chapter 4, we know that the equivalent results can be obtained using spatial convolution.

Note that the term inside the brackets in

$$f(x, y) = \int_0^\pi \left[ \int_{-\infty}^\infty |\omega| G(\omega, \theta) e^{j2\pi\omega\rho} d\omega \right]_{\rho=x\cos\theta+y\sin\theta} d\theta \quad (5.11-16)$$

is the inverse Fourier transform of the product of two frequency domain functions. According to the convolution theorem, they are equal to the convolution of the spatial representations (inverse Fourier transform) of these two functions.

Let  $s(\rho)$  denote the inverse Fourier transform of  $|\omega|$ , we can write (5.11-16) as



$$\begin{aligned}
 f(x, y) &= \int_0^\pi \left[ \int_{-\infty}^\infty |\omega| G(\omega, \theta) e^{j2\pi\omega\rho} d\omega \right]_{\rho=x\cos\theta+y\sin\theta} d\theta \\
 &= \int_0^\pi [s(\rho) \star g(\rho, \theta)]_{\rho=x\cos\theta+y\sin\theta} d\theta \\
 &= \int_0^\pi \left[ \int_{-\infty}^\infty g(\rho, \theta) s(x\cos\theta + y\sin\theta - \rho) d\rho \right] d\theta
 \end{aligned} \tag{5.11-18}$$

The last two lines of (5.11-18) say the same thing: Individual **backprojections** at an angle  $\theta$  can be obtained by **convolving** the corresponding projection,  $g(\rho, \theta)$ , and the **inverse Fourier transform** of the **ramp filter**,  $s(\rho)$ .

With the exception of round off differences in computation, the results of using **convolution** will be identical to the results using **FFT**.

In general, **convolution** turns out to be more computationally efficient and is used in most of modern **CT** systems, while **Fourier transform** plays a central role in **theoretical** formulations and **algorithm** development.

### 5.11.6 Reconstruction Using Fan-beam Filtered Backprojections